

LECTURE BLOCK 3C: FOURIER SERIES OF PERIODIC SIGNALS

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Fourier Analysis Techniques

We will learn:

- Express periodic signals in term of Fourier Series.
- Use Fourier Series to analyse system driven by continuous periodic signals.
- Apply Parseval's theorem to compute the power or energy contained in a signal.
- Study Fourier Series' examples, and properties,
- Look at Discrete Time Fourier Series.

Discrete Time Fourier Series (DTFS)

Fourier Series expansions of **discrete-time periodic signals**;
But, DTFS is finite in length (has a finite number of terms).

Computation of its expansion coefficients requires **summations rather than integrals**. The response of an LTI system to a discrete-time periodic signal can be determined by applying the **superposition principle**. The process entails the following steps:

Step 1: Compute the DTFS of the input signal.

Step 2: Compute the output response to each DTFS term using the system's frequency response $\mathbf{H}(e^{j\Omega})$, as defined by
$$\mathbf{H}(e^{j\Omega}) = \sum_{i=0}^{\infty} h[i] e^{-ji\Omega} \quad .$$

Step 3: Sum the results to obtain the total output signal.

Discrete vs. Cont. Time: Fourier Series

Discrete Time:

Continuous Time:

Periodic:

$$x[n] = x[n + N_0]$$

$$x(t) = x(t + T_0)$$

Fourier Series:

$$x[n] = \sum_{k=0}^{N_0-1} \mathbf{x}_k e^{jk\Omega_0 n}$$

$$x(t) = \sum_{n=-\infty}^{\infty} \mathbf{x}_n e^{jn\omega_0 t}$$

Computing Coefficient:

$$\mathbf{x}_k = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x[n] e^{-jk\Omega_0 n}$$

$$\mathbf{x}_n = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-jn\omega_0 t} dt.$$

Terms:

$$N_0$$

Infinity, in general

A few notes:

► A periodic signal $x[n]$ of period N_0 is represented in its DTFS by the sum of N_0 complex exponentials at harmonic values of $\Omega_0 = 2\pi/N_0$. ◀

In general, the expansion coefficients are complex quantities,

$$\mathbf{x}_k = |\mathbf{x}_k|e^{j\phi_k}.$$

► Plots of $|\mathbf{x}_k|$ and ϕ_k as a function of k from $k = 0$ to $k = N_0 - 1$ constitute *magnitude and phase line-spectra*. ◀

Orthogonality property

$$\sum_{n=0}^{N_0-1} e^{j2\pi[(k-m)/N_0]n} = N_0\delta[k-m]$$
$$= \begin{cases} N_0 & \text{if } k = m, \\ 0 & \text{if } k \neq m, \end{cases}$$

where k , m , and n are integers, and none are larger than N_0 .

(a) $k = m$:

$$\sum_{n=0}^{N_0-1} e^0 = \sum_{n=0}^{N_0-1} 1 = N_0 \quad (k = m).$$

(b) $k \neq m$:

If in the finite geometric series

$$\sum_{n=0}^{N_0-1} \mathbf{r}^n = \frac{\mathbf{r}^{N_0} - 1}{\mathbf{r} - 1} \quad \text{for } \mathbf{r} \neq 1,$$

we replace $\mathbf{r}$ with $e^{j2\pi(k-m)/N_0}$, we have

$$\sum_{n=0}^{N_0-1} e^{j2\pi[(k-m)/N_0]n} = \frac{e^{j2\pi[(k-m)/N_0]N_0} - 1}{e^{j2\pi(k-m)/N_0} - 1}.$$

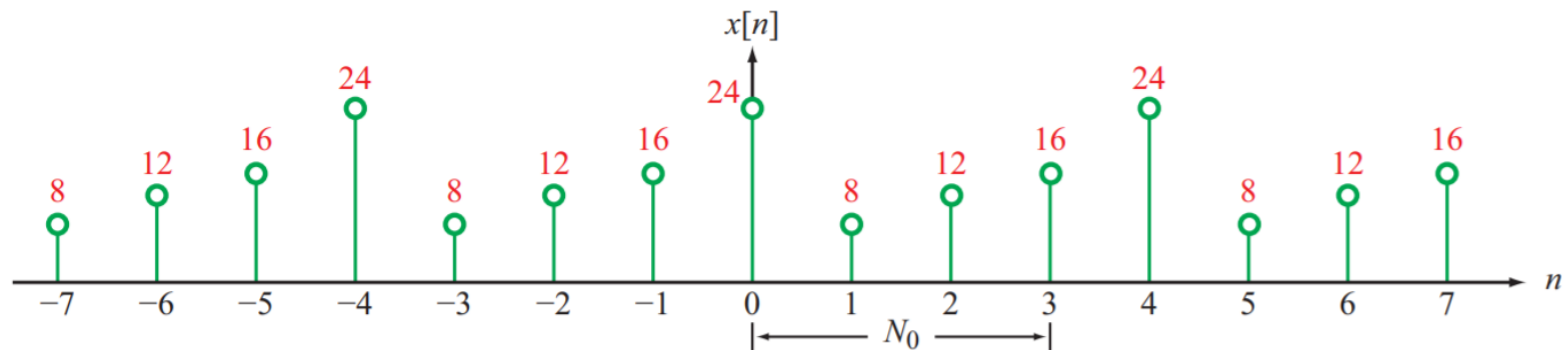
Since $k \neq m$, and k and m cannot be larger than N_0 , the exponent $j2\pi(k-m)/N_0$ in the denominator cannot be a multiple of 2π . Hence, the denominator cannot be zero. The exponent in the numerator, on the other hand, is $j2\pi(k-m)$, and therefore, it is always a multiple of 2π . Consequently,

$$\sum_{n=0}^{N_0-1} e^{j2\pi[(k-m)/N_0]n} = 0 \quad (k \neq m).$$



Example 1: DTFS Computation

Compute the DTFS of the periodic signal
 $x[n] = \{\dots, \underline{24}, 8, 12, 16, 24, 8, 12, 16, \dots\}$.



(a) $x[n]$

Example 1: DTFS Computation

Solution: The period of $x[n]$ is 4. So $N_0 = 4$
Plugging into $\mathbf{x}_k = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x[n] e^{-jk\Omega_0 n}$ gives

$$\mathbf{x}_0 = \frac{1}{4} (24 + 8 + 12 + 16) = 15.$$

$$\mathbf{x}_1 = \frac{1}{4} (24e^0 + 8e^{-j\pi/2} + 12e^{-j\pi} + 16e^{-j3\pi/2}) = 3 + j2 = 3.6e^{j33.7^\circ}.$$

$$\mathbf{x}_2 = \frac{1}{4} (24 - 8 + 12 - 16) = 3.$$

$$\mathbf{x}_3 = \mathbf{x}_1^* = 3 - j2 = 3.6e^{-j33.7^\circ}.$$

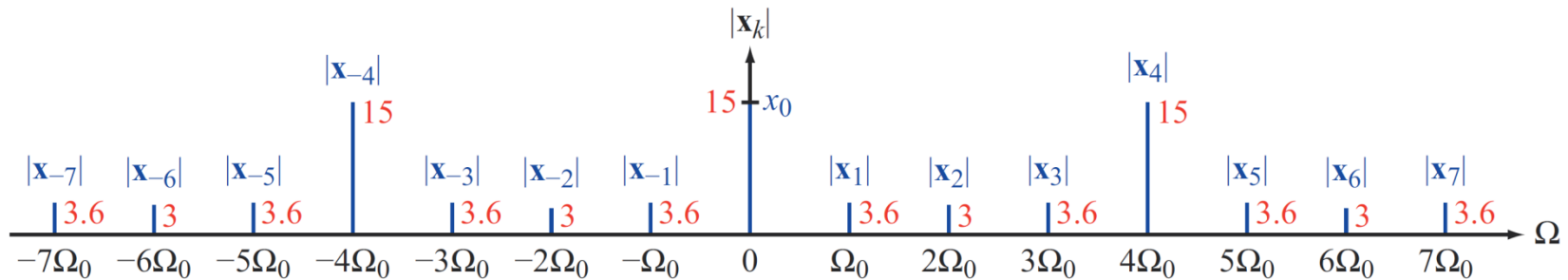
DTFS Example, Continued

The DTFS expansions of the given periodic signal $x[n]$ is then

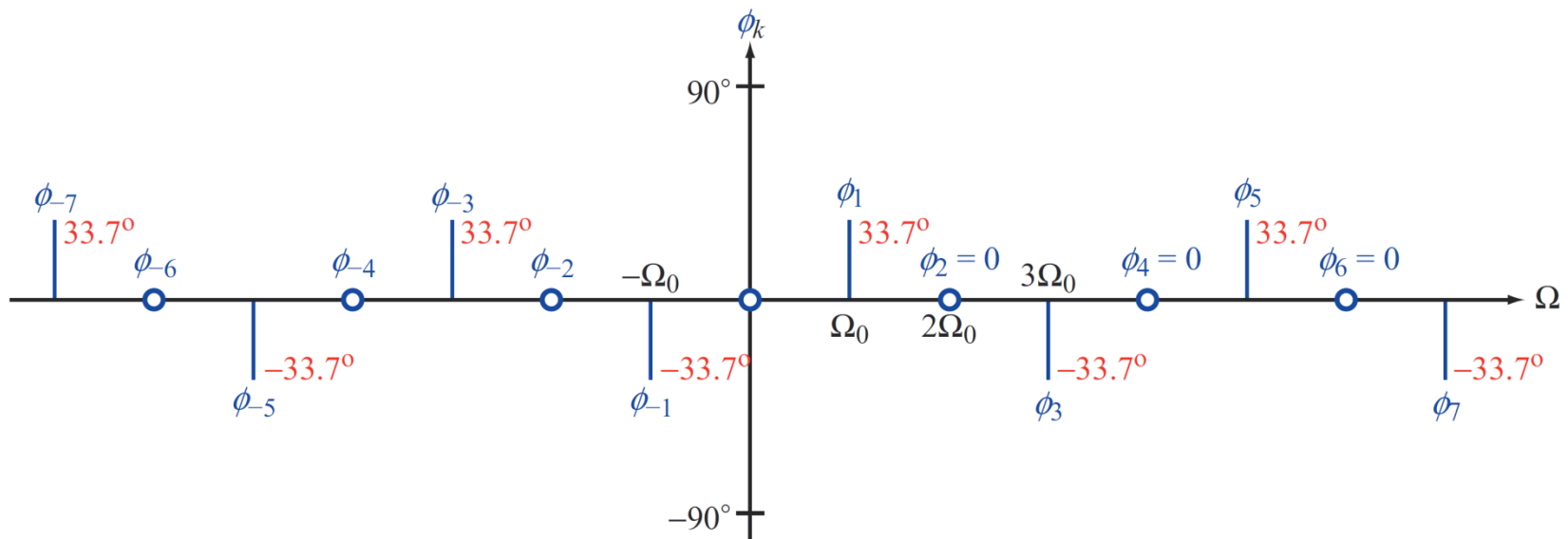
$$\begin{aligned}x[n] &= x_0 + \mathbf{x}_1 e^{j\pi n/2} + \mathbf{x}_2 e^{j\pi n} + \mathbf{x}_3 e^{j3\pi n/2} \\&= 15 + 3.6e^{j33.7^\circ} e^{j\pi n/2} + 3e^{j\pi n} + 3.6e^{-j33.7^\circ} e^{j3\pi n/2}. \\&= 15 + 7.2 \cos\left(\frac{\pi n}{2} + 33.7^\circ\right) + 3 \cos(\pi n).\end{aligned}$$

Magnitude and phase **line spectra** of $x[n]$ on next slide.

DTFS Line Spectra for Example



(b) Magnitude line spectrum



(c) Phase line spectrum

Spectral Symmetry

- In Example 1, were we to use
$$\mathbf{x}_k = \frac{1}{N_0} \sum_{n=0}^{N_0-1} x[n] e^{-jk\Omega_0 n}$$
 to compute $\mathbf{x}_k$ for negative values of k , the outcome would have been $\mathbf{x}_{-k} = \mathbf{x}_k^*$,

which means that the **magnitude line spectrum is an even function** and the **phase line spectrum is an odd function**. This is consistent with CTFS.

- Were we to compute $\mathbf{x}_k$ for values of k greater than 3, we would discover that the line spectra repeats periodically at a period of four lines. This is very different from the spectra of periodic continuous-time signals.

So we notice that:

- (1) Both spectra repeat every $N_0 = 4$ lines.
- (2) Magnitude spectrum $|\mathbf{x}_k|$ is an even function.
- (3) Phase spectrum ϕ_k is an odd function.

DTFS: Parseval's Theorem

Average power is the same
In time or frequency domain

$$\frac{1}{N_0} \sum_{n=0}^{N_0-1} |x[n]|^2 = \sum_{k=0}^{N_0-1} |\mathbf{x}_k|^2.$$

For the previous example, we have:

Time Domain: $P_{\text{av}} = \frac{1}{4} \left[(24)^2 + (8)^2 + (12)^2 + (16)^2 \right] = 260.$

Frequency Domain: $P_{\text{av}} = (15)^2 + (3.6)^2 + 3^2 + (3.6)^2 = 260.$

(Note that 3.6 is actually $\sqrt{13}$ rounded off.)

TABLE 3.1 PROPERTIES OF CONTINUOUS-TIME FOURIER SERIES

Property	Section	Periodic Signal	Fourier Series Coefficients
		$\left. \begin{array}{l} x(t) \\ y(t) \end{array} \right\} \begin{array}{l} \text{Periodic with period } T \text{ and} \\ \text{fundamental frequency } \omega_0 = 2\pi/T \end{array}$	$\begin{array}{l} a_k \\ b_k \end{array}$
Linearity	3.5.1	$Ax(t) + By(t)$	$Aa_k + Bb_k$
Time Shifting	3.5.2	$x(t - t_0)$	$a_k e^{-jk\omega_0 t_0} = a_k e^{-jk(2\pi/T)t_0}$
Frequency Shifting		$e^{jM\omega_0 t} x(t) = e^{jM(2\pi/T)t} x(t)$	a_{k-M}
Conjugation	3.5.6	$x^*(t)$	a_{-k}^*
Time Reversal	3.5.3	$x(-t)$	a_{-k}
Time Scaling	3.5.4	$x(\alpha t), \alpha > 0$ (periodic with period T/α)	a_k
Periodic Convolution		$\int_T x(\tau)y(t - \tau)d\tau$	$Ta_k b_k$
Multiplication	3.5.5	$x(t)y(t)$	$\sum_{l=-\infty}^{+\infty} a_l b_{k-l}$
Differentiation		$\frac{dx(t)}{dt}$	$jk\omega_0 a_k = jk\frac{2\pi}{T} a_k$
Integration		$\int_{-\infty}^t x(t) dt$ (finite valued and periodic only if $a_0 = 0$)	$\left(\frac{1}{jk\omega_0}\right)a_k = \left(\frac{1}{jk(2\pi/T)}\right)a_k$
Conjugate Symmetry for Real Signals	3.5.6	$x(t)$ real	$\begin{cases} a_k = a_{-k}^* \\ \Re\{a_k\} = \Re\{a_{-k}\} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \\ a_k = a_{-k} \\ \angle a_k = -\angle a_{-k} \end{cases}$
Real and Even Signals	3.5.6	$x(t)$ real and even	a_k real and even
Real and Odd Signals	3.5.6	$x(t)$ real and odd	a_k purely imaginary and odd
Even-Odd Decomposition of Real Signals		$\begin{cases} x_e(t) = \mathcal{E}\{x(t)\} & [x(t) \text{ real}] \\ x_o(t) = \mathcal{O}\{x(t)\} & [x(t) \text{ real}] \end{cases}$	$\begin{array}{l} \Re\{a_k\} \\ j\Im\{a_k\} \end{array}$
Parseval's Relation for Periodic Signals			
$\frac{1}{T} \int_T x(t) ^2 dt = \sum_{k=-\infty}^{+\infty} a_k ^2$			

TABLE 3.2 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

Property	Periodic Signal	Fourier Series Coefficients
	$\left. \begin{array}{l} x[n] \\ y[n] \end{array} \right\} \begin{array}{l} \text{Periodic with period } N \text{ and} \\ \text{fundamental frequency } \omega_0 = 2\pi/N \end{array}$	$\left. \begin{array}{l} a_k \\ b_k \end{array} \right\} \begin{array}{l} \text{Periodic with} \\ \text{period } N \end{array}$
Linearity	$Ax[n] + By[n]$	$Aa_k + Bb_k$
Time Shifting	$x[n - n_0]$	$a_k e^{-jk(2\pi/N)n_0}$
Frequency Shifting	$e^{jM(2\pi/N)n} x[n]$	a_{k-M}
Conjugation	$x^*[n]$	a_{-k}^*
Time Reversal	$x[-n]$	a_{-k}
Time Scaling	$x_{(m)}[n] = \begin{cases} x[n/m], & \text{if } n \text{ is a multiple of } m \\ 0, & \text{if } n \text{ is not a multiple of } m \end{cases}$ (periodic with period mN)	$\frac{1}{m} a_k$ (viewed as periodic with period mN)
Periodic Convolution	$\sum_{r=\langle N \rangle} x[r]y[n-r]$	$Na_k b_k$
Multiplication	$x[n]y[n]$	$\sum_{l=\langle N \rangle} a_l b_{k-l}$
First Difference	$x[n] - x[n-1]$	$(1 - e^{-jk(2\pi/N)})a_k$
Running Sum	$\sum_{k=-\infty}^n x[k] \quad \left(\begin{array}{l} \text{finite valued and periodic only} \\ \text{if } a_0 = 0 \end{array} \right)$	$\left(\frac{1}{(1 - e^{-jk(2\pi/N)})} \right) a_k$
Conjugate Symmetry for Real Signals	$x[n]$ real	$\begin{cases} a_k = a_{-k}^* \\ \Re\{a_k\} = \Re\{a_{-k}\} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \\ a_k = a_{-k} \\ \angle a_k = -\angle a_{-k} \end{cases}$
Real and Even Signals	$x[n]$ real and even	a_k real and even
Real and Odd Signals	$x[n]$ real and odd	a_k purely imaginary and odd
Even-Odd Decomposition of Real Signals	$\begin{cases} x_e[n] = \mathcal{E}v\{x[n]\} & [x[n] \text{ real}] \\ x_o[n] = \mathcal{O}d\{x[n]\} & [x[n] \text{ real}] \end{cases}$	$\begin{cases} \Re\{a_k\} \\ j\Im\{a_k\} \end{cases}$
Parseval's Relation for Periodic Signals		
$\frac{1}{N} \sum_{n=\langle N \rangle} x[n] ^2 = \sum_{k=\langle N \rangle} a_k ^2$		